

5.3.6 Time Reversal

$$Y(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} y[n]e^{-j\omega n} = \sum_{n=-\infty}^{+\infty} x[-n]e^{-j\omega n}$$

$$Y(e^{j\omega}) = \sum_{m=-\infty}^{+\infty} x[m]e^{-j(-\omega)m} = X(e^{-j\omega})$$

$$x[-n] \xleftrightarrow{\mathfrak{T}} X(e^{-j\omega})$$

5.3.7 Time Expansion

For Continuous-time signals

$$x(at) \xleftrightarrow{\mathfrak{T}} \frac{1}{|a|} X\left(\frac{j\omega}{a}\right)$$

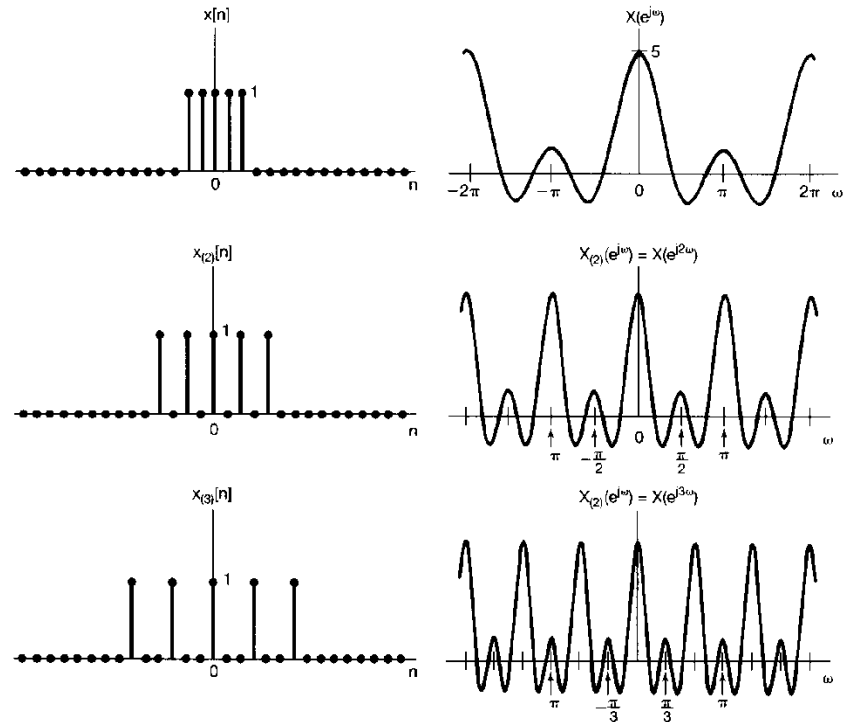
It is different for discrete-time signal.

$$x_{(k)}[n] = \begin{cases} x[n/k], & \text{if } n \text{ is a multiple of } k \\ 0, & \text{if } n \text{ is not a multiple of } k \end{cases}$$

$$x_{(k)}[n] \xleftrightarrow{\mathfrak{F}} X(e^{jk\omega})$$

Proof)

$$\begin{aligned} X_{(k)}(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} x_{(k)}[n] e^{-j\omega n} \\ n = \begin{cases} rk \\ rk+1, rk+2, \dots, rk+(k-1) \end{cases} & \left(\begin{aligned} &= \sum_{r=-\infty}^{\infty} x_{(k)}[rk] e^{-j\omega rk} \\ &= \sum_{r=-\infty}^{\infty} x[r] e^{-j\omega rk} \\ &= \sum_{r=-\infty}^{\infty} x[r] e^{-j(k\omega)r} \end{aligned} \right. \end{aligned}$$



5.3.8 Differentiation in frequency

$$nx[n] \xleftrightarrow{\mathfrak{F}} j \frac{dX(e^{j\omega})}{d\omega}$$

5.3.9 Parseval's relation

$$\sum_{n=-\infty}^{+\infty} |x[n]|^2 = \frac{1}{2\pi} \int_{2\pi} |X(e^{j\omega})|^2 d\omega$$

$|X(e^{j\omega})|^2$; the energy density spectrum of $x[n]$

Ex. 5.10)

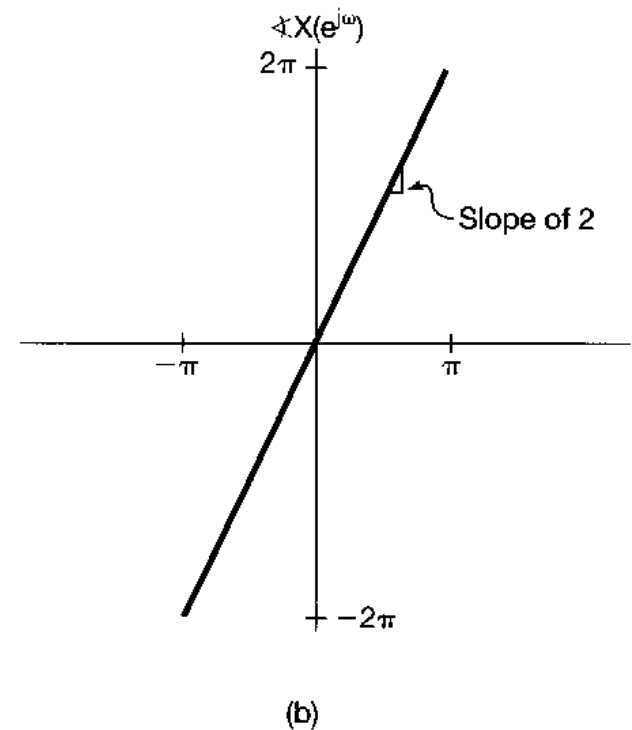
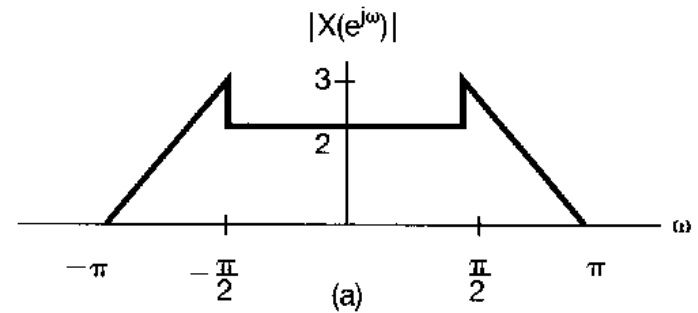
Is the time-domain signal $x[n]$

Periodic?

Real?

Even?

Of finite energy?



5.4 The Convolution Property

$$y[n] = x[n] * h[n] \quad \longleftrightarrow \quad Y(e^{j\omega}) = X(e^{j\omega})H(e^{j\omega})$$

Example : convolution of two sequences

$$x[n] = \{3, 2, 1\}, \quad h[n] = \{1, 2, 1\} \quad y[n] = x[n] * h[n] = \sum_{m=-\infty}^{\infty} x[m]h[n-m]$$

$$X(e^{j\omega}) = 3 + 2e^{-j\omega} + e^{-j2\omega}$$

$$H(e^{j\omega}) = 1 + 2e^{-j\omega} + e^{-j2\omega}$$

$$Y(e^{j\omega}) = X(e^{j\omega})H(e^{j\omega})$$

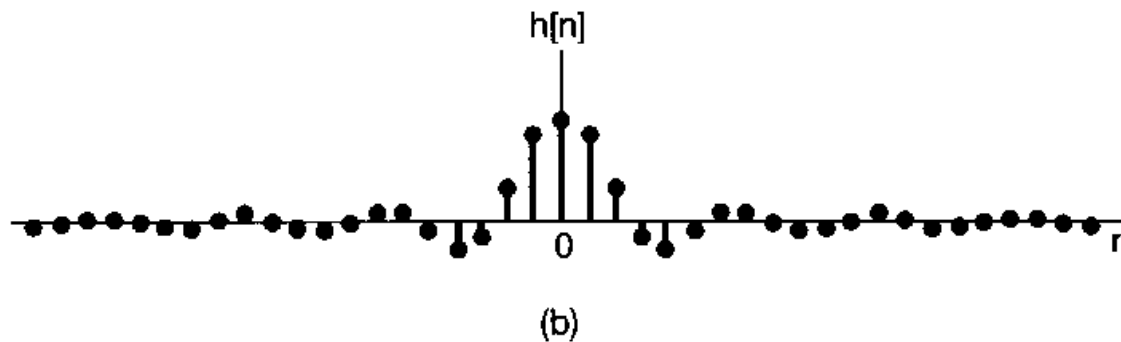
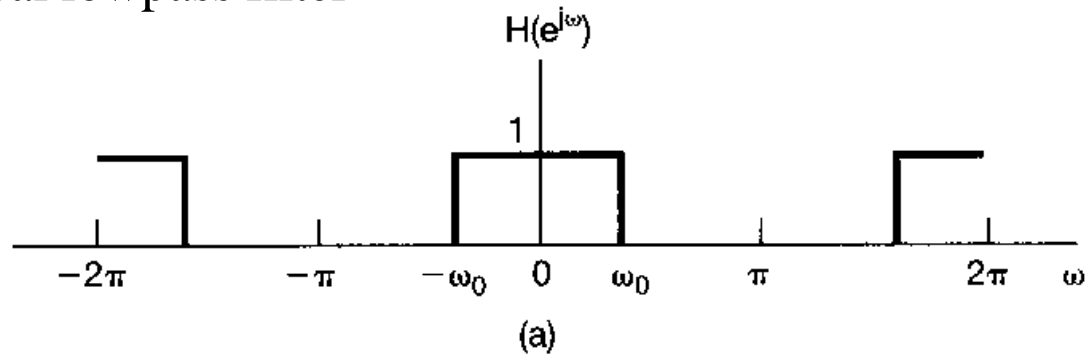
$$= 3 + (6 + 2)e^{-j\omega} + (3 + 4 + 1)e^{-j2\omega} + (2 + 2)e^{-j3\omega} + e^{-j4\omega}$$

Ex. 5.11)

$$h[n] = \delta[n - n_0]$$

$$Y(e^{j\omega}) = e^{-j\omega n_0} X(e^{j\omega})$$

Ex. 5.12) Ideal lowpass filter



$$h[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega = \frac{1}{2\pi} \int_{-\omega_0}^{\omega_0} e^{j\omega n} d\omega = \frac{\sin \omega_0 n}{\pi n}$$

Ex. 5.13)

$$H(e^{j\omega}) = \frac{1}{1 - \alpha e^{-j\omega}} \qquad X(e^{j\omega}) = \frac{1}{1 - \beta e^{-j\omega}}$$

$$Y(e^{j\omega}) = H(e^{j\omega})X(e^{j\omega}) = \frac{1}{(1 - \alpha e^{-j\omega})(1 - \beta e^{-j\omega})}$$

$$Y(e^{j\omega}) = \frac{A}{(1 - \alpha e^{-j\omega})} + \frac{B}{(1 - \beta e^{-j\omega})}$$

$$\begin{aligned} y[n] &= \frac{\alpha}{\alpha - \beta} \alpha^n u[n] - \frac{\beta}{\alpha - \beta} \beta^n u[n] \\ &= \frac{1}{\alpha - \beta} [\alpha^{n+1} u[n] - \beta^{n+1} u[n]] \end{aligned}$$

For $\alpha = \beta$,
$$Y(e^{j\omega}) = \left(\frac{1}{1 - \alpha e^{-j\omega}} \right)^2$$

$$Y(e^{j\omega}) = \frac{j}{\alpha} e^{j\omega} \frac{d}{d\omega} \left(\frac{1}{1 - \alpha e^{-j\omega}} \right)$$

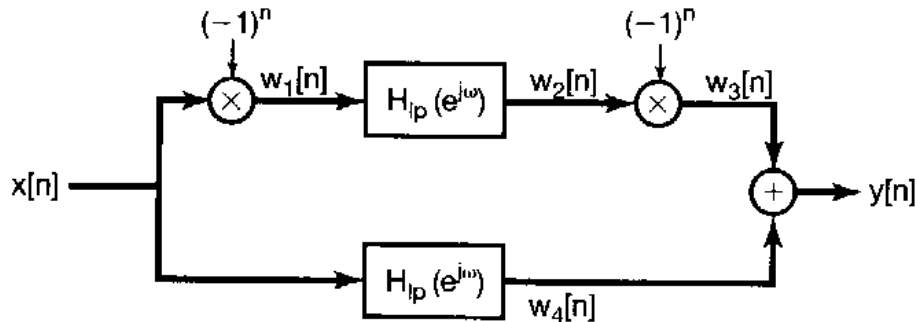
$$\alpha^n u[n] \xleftrightarrow{\mathfrak{F}} \left(\frac{1}{1 - \alpha e^{-j\omega}} \right)$$

$$n\alpha^n u[n] \xleftrightarrow{\mathfrak{F}} j \frac{d}{d\omega} \left(\frac{1}{1 - \alpha e^{-j\omega}} \right) \quad (\text{Differentiation in frequency})$$

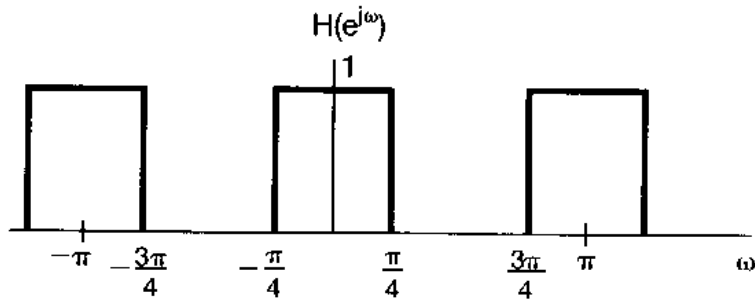
$$(n+1)\alpha^{n+1} u[n+1] \xleftrightarrow{\mathfrak{F}} j e^{j\omega} \frac{d}{d\omega} \left(\frac{1}{1 - \alpha e^{-j\omega}} \right)$$

$$\therefore y[n] = (n+1)\alpha^n u[n+1] = (n+1)\alpha^n u[n]$$

Ex. 5.14) Ideal bandstop filter at $\frac{\pi}{4} < |\omega| < \frac{3\pi}{4}$



(a)



(b)

$$W_1(e^{j\omega}) = X(e^{j(\omega-\pi)})$$

$$W_2(e^{j\omega}) = H_{lp}(e^{j\omega})X(e^{j(\omega-\pi)})$$

$$\begin{aligned} W_3(e^{j\omega}) &= H_{lp}(e^{j(\omega-\pi)})X(e^{j(\omega-2\pi)}) \\ &= H_{lp}(e^{j(\omega-\pi)})X(e^{j\omega}) \end{aligned}$$

$$W_4(e^{j\omega}) = H_{lp}(e^{j\omega})X(e^{j\omega})$$

$$\begin{aligned} Y(e^{j\omega}) &= W_3(e^{j\omega}) + W_4(e^{j\omega}) \\ &= [H_{lp}(e^{j(\omega-\pi)}) + H_{lp}(e^{j\omega})]X(e^{j\omega}) \end{aligned}$$

$$\therefore H(e^{j\omega}) = H_{lp}(e^{j(\omega-\pi)}) + H_{lp}(e^{j\omega})$$

5.5 The multiplication property

$$y[n] = x_1[n]x_2[n]$$

$$\begin{aligned} Y(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} y[n]e^{-j\omega n} = \sum_{n=-\infty}^{\infty} x_1[n]x_2[n]e^{-j\omega n} \\ &= \sum_{n=-\infty}^{\infty} x_2[n] \left\{ \frac{1}{2\pi} \int_{2\pi} X_1(e^{j\theta}) e^{j\theta n} d\theta \right\} e^{-j\omega n} \\ &= \frac{1}{2\pi} \left\{ \int_{2\pi} X_1(e^{j\theta}) \sum_{n=-\infty}^{+\infty} x_2[n] e^{-j(\omega-\theta)n} d\theta \right\} \end{aligned}$$

$$Y(e^{j\omega}) = \frac{1}{2\pi} \left\{ \int_{2\pi} X_1(e^{j\theta}) X_2(e^{j(\omega-\theta)}) d\theta \right\}$$

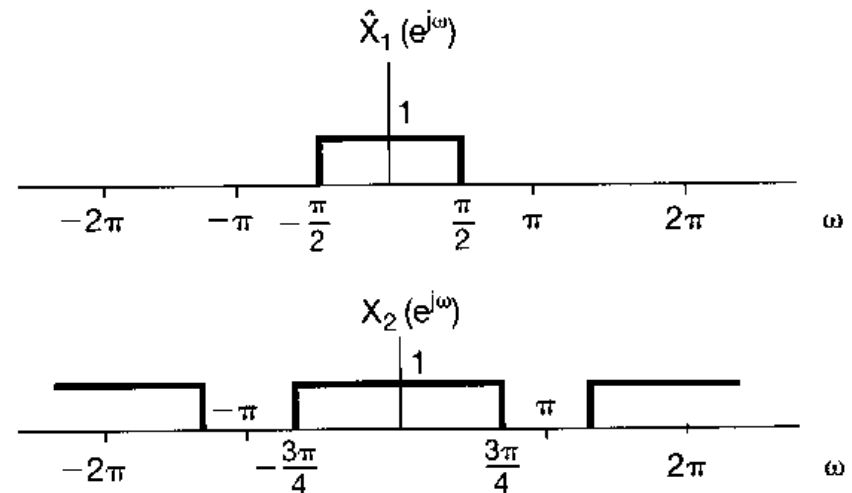
Ex. 5.15)

$$x[n] = x_1[n]x_2[n], \quad x_1[n] = \frac{\sin(\pi n/2)}{\pi n}, \quad x_2[n] = \frac{\sin(3\pi n/4)}{\pi n} \quad (\text{Text Correction})$$

$$X(e^{j\omega}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X_1(e^{j\theta})X_2(e^{j(\omega-\theta)})d\theta$$

$$\hat{X}_1(e^{j\omega}) = \begin{cases} X_1(e^{j\omega}) & \text{for } -\pi < \omega \leq \pi \\ 0 & \text{otherwise} \end{cases}$$

$$\begin{aligned} X(e^{j\omega}) &= \frac{1}{2\pi} \int_{-\pi}^{\pi} \hat{X}_1(e^{j\theta})X_2(e^{j(\omega-\theta)})d\theta \\ &= \frac{1}{2\pi} \int_{-\infty}^{\infty} \hat{X}_1(e^{j\theta})X_2(e^{j(\omega-\theta)})d\theta \end{aligned}$$



Result: Fig. 5.20, page 390

5.7 Duality

5.7.1 In the D.T. Fourier Series

$$x[n] \xleftrightarrow{\mathcal{F}S} a_k$$

$$a_n \xleftrightarrow{\mathcal{F}S} \frac{1}{N} x[-k]$$

$$x[n] = \sum_{k=\langle N \rangle} a_k e^{j\frac{2\pi}{N}kn}$$

$$a_k = \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-j\frac{2\pi}{N}kn} \xrightarrow{\begin{matrix} k \rightarrow n \\ (-n) \rightarrow k \end{matrix}} a_n = \frac{1}{N} \sum_{k=\langle N \rangle} x[-k] e^{j\frac{2\pi}{N}kn}$$

5.7.2 Between the DTFT and CTFS

$$X(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} x[n] e^{-j\omega n} = \sum_{m=-\infty}^{+\infty} x[-m] e^{j\omega m}$$

$$x(t) = \sum_{k=-\infty}^{+\infty} a_k e^{jk\omega_0 t}$$

\Rightarrow Fourier series representation of $X(e^{j\omega})$

$$X(e^{jt}) = \sum_{m=-\infty}^{+\infty} x[-m] e^{jmt} = \sum_{m=-\infty}^{+\infty} x[-m] e^{jm\omega_0 t} \quad \left(\because \omega_0 = \frac{2\pi}{T_0} = \frac{2\pi}{2\pi} = 1 \right)$$

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

$$a_k = \frac{1}{T} \int_T x(t) e^{-jk\omega_0 t} dt$$



5.8 Systems characterized by linear constant-coefficient difference equation

$$y[n] - ay[n-1] = x[n]$$

$$H(e^{j\omega}) = \frac{1}{1 - ae^{-j\omega}} \quad h[n] = a^n u[n]$$

$$y[n] - \frac{3}{4}y[n-1] + \frac{1}{8}y[n-2] = 2x[n]$$

$$\begin{aligned}
 H(e^{j\omega}) &= \frac{2}{1 - \frac{3}{4}e^{-j\omega} + \frac{1}{8}e^{-j2\omega}} \\
 &= \frac{2}{\left(1 - \frac{1}{2}e^{-j\omega}\right)\left(1 - \frac{1}{4}e^{-j\omega}\right)} = \frac{4}{\left(1 - \frac{1}{2}e^{-j\omega}\right)} - \frac{2}{\left(1 - \frac{1}{4}e^{-j\omega}\right)}
 \end{aligned}$$

$$h[n] = 4\left(\frac{1}{2}\right)^n u[n] - 2\left(\frac{1}{4}\right)^n u[n]$$

Ex. 5.20)

$$Y(e^{j\omega}) = H(e^{j\omega})X(e^{j\omega}) = \left[\frac{2}{\left(1 - \frac{1}{2}e^{-j\omega}\right)\left(1 - \frac{1}{4}e^{-j\omega}\right)} \right] \left[\frac{1}{\left(1 - \frac{1}{4}e^{-j\omega}\right)} \right]$$
$$= \frac{2}{\left(1 - \frac{1}{2}e^{-j\omega}\right)\left(1 - \frac{1}{4}e^{-j\omega}\right)^2}$$

$$Y(e^{j\omega}) = \frac{B_{11}}{\left(1 - \frac{1}{4}e^{-j\omega}\right)} + \frac{B_{12}}{\left(1 - \frac{1}{4}e^{-j\omega}\right)^2} + \frac{B_{21}}{\left(1 - \frac{1}{2}e^{-j\omega}\right)}$$

$$Y(e^{j\omega}) = -\frac{4}{\left(1 - \frac{1}{4}e^{-j\omega}\right)} - \frac{2}{\left(1 - \frac{1}{4}e^{-j\omega}\right)^2} + \frac{8}{\left(1 - \frac{1}{2}e^{-j\omega}\right)}$$

$$y[n] = \left\{ -4\left(\frac{1}{4}\right)^n - 2(n+1)\left(\frac{1}{4}\right)^n + 8\left(\frac{1}{2}\right)^n \right\} u[n]$$